

TRANSFORMER DESIGN SECTION-C

(21)

A T/I is a static electromagnetic device that transfers an a.c. electric power from one ckt to another at the same freq but the voltage level is changed.

T/I only serves a very important link b/w the generation of electrical energy and its utilization. The use of the T/I action is one of the major reason for

commercial preference of a.c. for power use. → an alt. ϕ is produced whose amplitude & no. of turns $E_p = 4.44 f \Phi m T_p$ \rightarrow Palt. volt source → two or more coils which links with common magnetic field.

Electrical energy is mostly generated at remote places where large water head is available for hydroelectric power stations or thermal energy in the form of coal for thermal power station. This energy is transmitted to long distances for use at load centres. Therefore the transmission and distribution of electrical energy is become economically possible by extensive use of transformers. In this area of applications the transformers are termed as power T/I and Distribution T/I's

$$E_s = 4.44 f \Phi m T_s \Rightarrow \frac{E_s}{E_p} = \frac{T_s}{T_p} \Rightarrow \text{Value of } E_s \text{ can be obtained}$$

DISTBⁿ T/I's:- up to size of about 200-250kVA used for stepping down the voltage to a standard service voltage.

They are continuously in the ckt whether they are carrying any load or not. The core losses would occurs for all the times whereas copper losses only when they are loaded. Therefore designer have to give attention to reduce core losses compared to full load copper losses and coils also arranging in such a way to ^{minimize} ~~reduce~~ the leakage reactance.

Power T/I's:→ Above 200-250kVA used in generating station or substn for transforming the voltage at each end of power transmission line.

they are put into operation during load hours (Q2) and thrown off during light load hours. These HTs are designed to have maximum efficiency at full load.

Some other types also like - instrument HT for metering purposes, [they are named as CT & PT]. In high voltage laboratory they are used to generate very high voltage for testing purpose termed as testing HT.

* Transformer type:- Acc. to construction

① Core & ② Shell type

Acc to service condition:-

↳ ① Distn HT &
② Power HT.

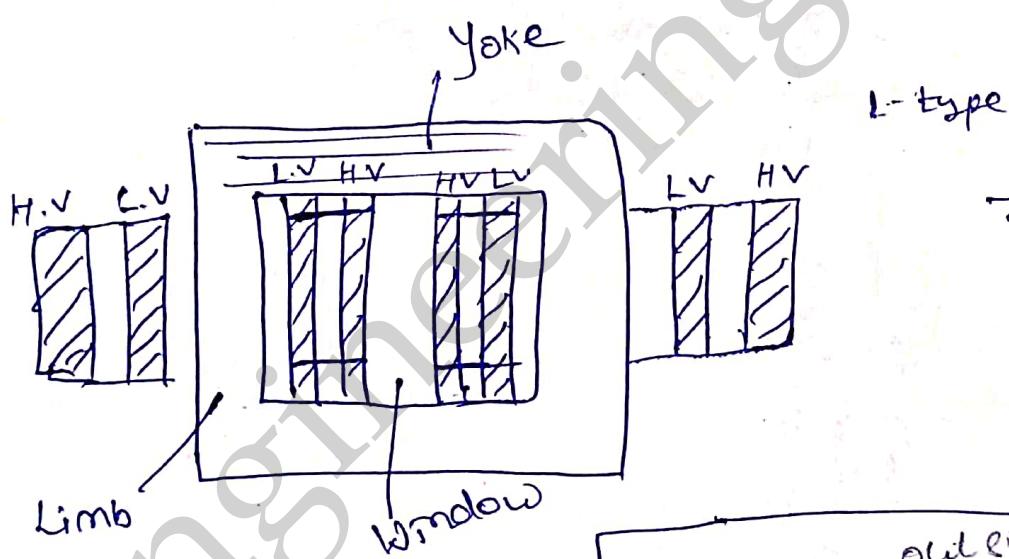
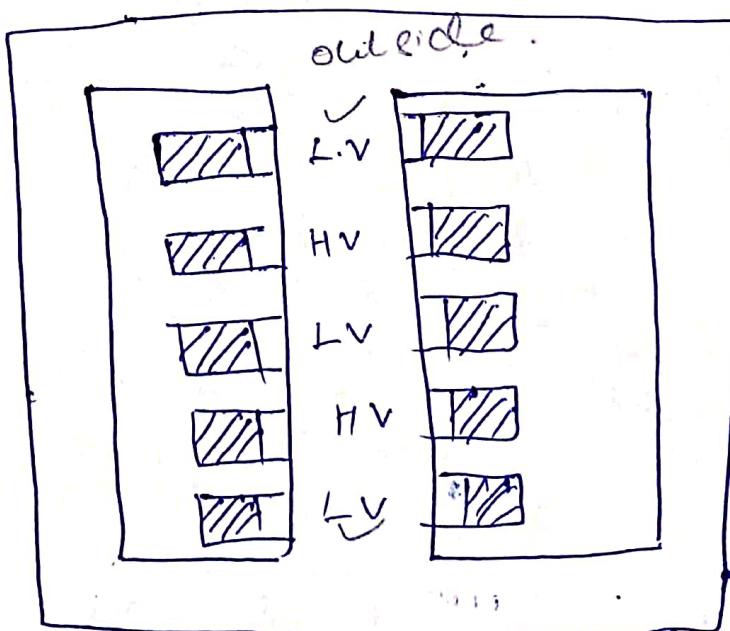


Fig-1 a Core type
concentric



both L.V & H.V windings are put around the central limb.

E-type

Fig 1 b Sandwiched

TYPES OF TRANSFORMERS

Conventional type
(With two wedge Dr. & sec.)

Q2

Conventional type

(With two wedge Dr. & sec.)

Auto transformers

(With one wedge.)

TRANSMISSION type

Distribution type

With Continuous Variation of V
230/V to 270V
(Variac)

Variation in steps
(Used as stabilizer)

EH type

HY type

Up to 66kV

Above 66kV

Medium Voltage
11kV/10.5kV

High Voltage
for checking 7
Insulation / Apparatus
upto 5000V.

Instrument
transformer

High Current
upto 200-400

CT
Current Transformer

Low Current
upto 100-150

PT
Potential Transformer

100/5A/200/5A/5A/5A, 1000/5

Millivoltmeter (over load)

Constructional parts: → fig page No-26

(23)

Static electric device having no rotating part.

- ① Magnetic ckt: - limb, yoke, and clamping structure
- ② Electric ckt: - Different windings
(primary, secondary (and tertiary, if any)
formers insulation.)
3. Terminals: - Leads and tappings, bushing, terminal insulators.
- ④ Cooling circuit: - tank, oil, conservator,
breath & auxiliary apparatus.

TI/T used in p.s applications are mainly single phase and 3-phase core type. Rarely used shell type (3-ph).
Here also we will mostly deal w/ & discuss with single phase core & shell type but 3-ph only core (no any 3-ph shell type).

* Windings: ✓ circular and rectangular
mostly used circular bcz rectangular ^{wdg} involves wastage of space and under short ckt condition it get deformed. Hence it may be employed only for very very small capacity T/T.
Now further circular are classified: -

- ① Concentric
- ② Sandwich



The concentric ~~cylindrical~~ ^{circular} wedges -
is further classified as: →

- ① Cylindrical
- ② Helical wedg.
- ③ Cross over wedg
- ④ Contineous Disc wedg.

* Specifications:-

1. Capacity i.e., rating in kVA.
2. Voltage rating, KV of primary & secondary sides.
3. No. of phases, 1-ph or 3-ph.
4. freq. in Hz.
5. Connections Y or Δ for 3-ph. HT.
6. Type - Core or shell (constructional pt. of view)
7. Type - power or Distribution → (Appn " ")
8. Type of cooling 9. Temperature rise, depending upon type of insulation.
10. Voltage regulation
11. No load current, 12. Efficiency 13. P.F.

* Output equation: → The o/p of HT in kVA is to be related with its main dimension. is called as o/p equation.

Let $O = \text{output, kVA}$

$\phi_m = \text{Main flux, wb}$ from fig page no. 26

$B_m = \text{Max. flux density A/m}^2$

$A_i = \text{Net cross section area of iron core, m}^2$, Net core area

$A_{gi} = \text{Gross (iron) core section, m}^2$, gross core area

$A_w = \text{Net window area, m}^2$ ($A_i = k_s A_{gi}$) (including stacking factor 0.85-0.9)

$D = \text{Dist. b/w core centres, m}$

$d = \text{Diam of circumscribing circle around core, m}$

$K_w = \text{Window space factor, f = freq, Hz}$

$E_t = \text{E.m.f per turn, volt}$

$T_1, T_2 = \text{No. of 1" \& 2" turns, I}_1, I_2 = 1" \& 2" \text{ currents}$

$V_1, V_2 = \text{Primary \& Secondary phase voltage, Volt.}$

$a_1, a_2 = \text{primary \& secondary conductor sections, m}^2$

$L_i = \text{mean length of flux path in iron, m}$

$L_m = \text{length of mean turn of wdg, m}$

G_{2i} = weight of active part of 1 m², kg

G_{2c} = " " copper, kg

ρ_i = specific core losses, W per kg

ρ_c = " copper " "

output E_m for

* 1-phase core type T/T: -

E.m.f. induced in T/T wdg as

$$E = 4.44 f \phi mT$$

The e.m.f per turn

$$E_t = \frac{E}{T} = 4.44 f \phi m \quad \text{--- --- --- ①}$$

whether the primary & secondary turn is the part of

that it Lnk with flux ϕ_m .

The window space factor k_w is defined as ratio of copper area in window to total window area. i.e,

$$k_w = \frac{\text{Copper area in Window}}{\text{Total window area} (= A_w)}$$

Hence Copper area in window = $k_w A_w$ - ②

→ The copper area can also be found out by the no. of turns of the primary and secondary winding and the conductor sections. The window here contains only one primary and one secondary Here we are considering one 1"2" one 2" wdg. (both L.V. near L.M.)

Copper area in window = $T_1 Q_1 + T_2 Q_2$

$$= T_1 \frac{S_1 I_1}{S_e} + T_2 \frac{I_2}{S} \quad \left. \begin{array}{l} \text{Assuming that the} \\ \text{current density } S \text{ is} \\ \text{same for both the wdg.} \\ A, A_2 \rightarrow 1"2" \text{ cond sects} \end{array} \right\}$$

$$= \frac{1}{S} (T_1 I_1 + T_2 I_2)$$

$$= \frac{2 I T}{S}$$

(3) $I, T, \approx I_2 T_2 = I T$
if we neglect magnetising AT/mmf.

from Eqn (2) & (3)

$$K_W A_W = \frac{2IT}{S} \Rightarrow IT = \frac{K_W A_W S}{2}$$

Now rating of 1-ph. T/T is $(Q = \epsilon \times I \times 10^{-3}) \rightarrow (4)$

$$Q = V_{ph} I_{ph} \times 10^{-3} \approx E_{ph} I_{ph} \times 10^{-3}$$

[As V_{ph} is approx equal to E_{ph}]

$$= (4.44 f \phi_m T) I \times 10^{-3}$$

$$= 4.44 f \phi_m (IT) \times 10^{-3}$$

$$= 4.44 f \phi_m \left[\frac{K_W A_W S}{2} \right] \times 10^{-3} \quad [\text{from Eqn (4)}]$$

$$= 2.22 f \phi_m K_W A_W S \times 10^{-3}$$

$$Q = 2.22 f B_m A_i A_W K_W S \times 10^{-3} \text{ kVA}$$

$$A_i = \frac{\phi_m}{B_m}$$

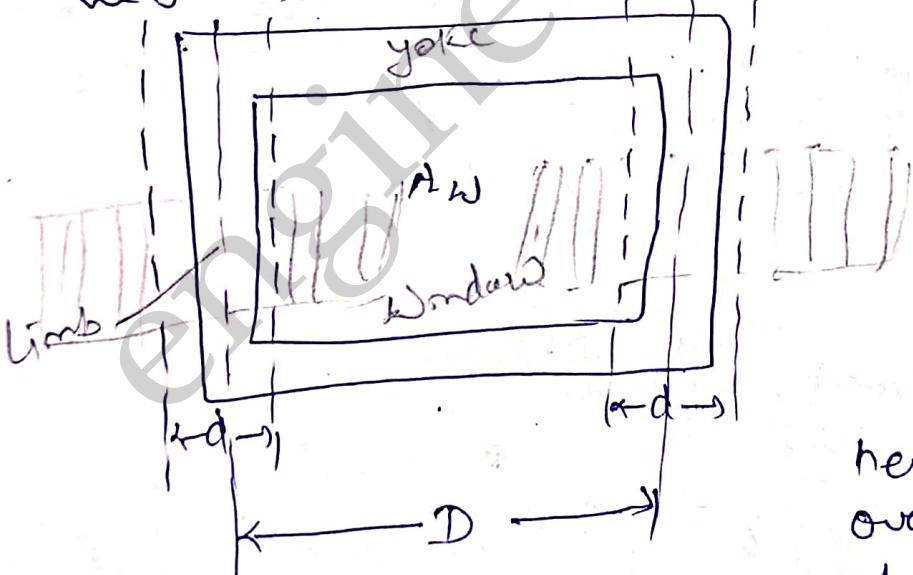
$$B_m = \frac{\phi}{A}$$

$$\text{As } [\phi_m = B_m A_i]$$

$B_m \rightarrow$ Max. Flux density, Wb/m^2

$A_i \rightarrow$ Net cross section area, m^2 .

Single phase shell type T/T:



1-ph. fig - shell type

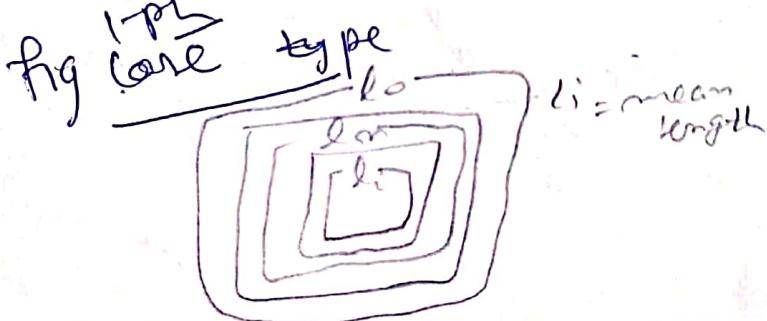


here both wedges are wound over the central limb. Each of window contains all the 1st & 2nd day tins.

Copper area in window

$$= T_1 a_1 + T_2 a_2 = \frac{2IT}{S}$$

$$= K_W A_W \text{ same}$$

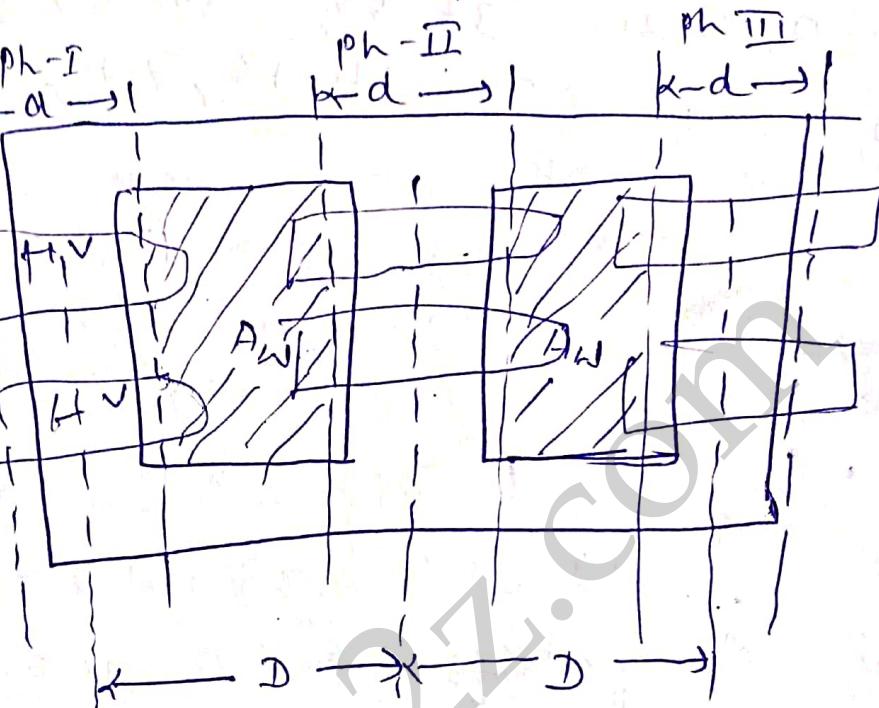


Three phase T/T : → (Core-type)

Each limb is wound with both primary and secondary wedges of respective phases.

Each window contains two primaries and two secondaries i.e. copper area in each window

$$= 2(T_1 a_1 + T_2 a_2)$$



$$= 2 \left[T_1 \frac{I_1}{S} + T_2 \frac{I_2}{S} \right] = \frac{2}{S} [I_1 T_1 + I_2 T_2]$$

$$= \frac{4 IT}{S} \quad [I_1 T_1 + I_2 T_2 = IT] \\ \text{If we neglect magnetic loss in AT/magnet}$$

(IT)

Also copper area in each window = $K_w A_w$

$$K_w A_w = \frac{4 IT}{S}$$

$$IT = \frac{K_w A_w S}{4} \quad \text{--- (6)}$$

Now ratings of 1-ph T/T is
Now output of 3-ph T/T $\{ S = 3 EI \times 10^{-3} \}$

$$\Phi = 3 \epsilon p \cdot I_{ph} \times 10^{-3} = 3 (4.44 f \phi_m T_{ph}) \cdot I_{ph} \times 10^{-3}$$

$$= 3 (4.44 f \phi_m) \left(\frac{K_w A_w S}{4} \right) \times 10^{-3} \quad \text{from (6)}$$

$$\boxed{\Phi = 3.33 f B_m A; A_w K_w S \times 10^{-3} \text{ kVA}} \quad [\phi_m = B_m \times A]$$

$$B_m = \frac{\Phi_m}{A_i}$$

Output Equation :- E.m.f / per turn at endg : $E_t = K \sqrt{\phi}$

volt/ per turn or

We have to relate the output kVA of T/f to e.m.f per turn that is the turning design factor for T/f. ~~from~~ from where a const value (K) we will obtain.

(Op) Rating for a 1-ph. T/f $\phi = \frac{mmf}{Res}$

$$\text{Rating} = E \times I \times 10^{-3}$$

$$= 4.44 f \phi m T \times I \times 10^{-3} \quad \text{--- (1)}$$

Any no. of Design should satisfy this Eq.

The flux (ϕ_m) is roughly a measure of the cross section of ~~a~~ iron core and (I_T , mmf) I_T gives the cross section of the winding.
Therefore the ratio of ϕ_m/I_T will be a const.

Let $\frac{\phi_m}{I_T} = \lambda$

$$(1) \Rightarrow \text{Rating} = 4.44 f \phi_m (I_T) \times 10^{-3}$$

$$= 4.44 f \frac{\phi_m^2}{\lambda} \times 10^{-3} \quad \left\{ I_T = \frac{\phi_m}{\lambda} \right\}$$

$$\phi_m^2 = \frac{\lambda \times 10^3}{4.44 f} \cdot Q$$

$$\phi_m = \sqrt{\frac{\lambda \times 10^3}{4.44 f}} \cdot \sqrt{\phi}$$

put this value $E = \frac{m \text{ ent}}{4.44 f \phi m T}$

$$E_t = E_t = 4.44 f \phi_m$$

$$= 4.44 f \left[\sqrt{\frac{\lambda \times 10^3}{4.44 f}} \sqrt{\phi} \right]$$

$$\begin{aligned} \text{mmf} &= \text{flux} \times \text{reluct} \\ \text{emf} &= I \times \text{Res.} \quad r' \\ \text{rel} &= \frac{\phi_m}{\text{mmf}} = \frac{\phi_m}{IT} \end{aligned}$$

$$= \lambda = \frac{1}{r'} \frac{\phi_m}{IT}$$

$$\frac{1}{\text{reluct}} = \frac{1}{r'} = \lambda$$

permeance

$$E_t = \sqrt{4.44 f \lambda \times 10^3} \sqrt{\phi}$$

$$\text{i.e., } E_t = K \sqrt{\phi} \quad \text{--- (2)}$$

where, $K = \sqrt{4.44 f \lambda \times 10^3}$

This K is a const and depends upon type, service condition and construction.

Table : P.T.O.

DESIGN OF CORE: →

Mainly the core used in the H/I are of rectangular and squared.

- for very small size of H/I - simple rectangular core section are used having circular or rectangular coils.
- As size of H/I are increased, we use generally the squared shape core and these squared core are - TWO STEPPED, THREE STEPPED and four stepped core sect'.

(i) SQUARE SECTION: → Referring fig-1

Gross core section
Gross area of core, $A_{gi} = a^2$

where, a = side of square and d = diameter of circumscribing circle

$$a = d / \sqrt{2} = 0.707d = 0.71d$$

$$\therefore A_{gi} = (0.71d)^2 = 0.5d^2$$

Net iron area / Net core section = $A_i = k_s A_{gi} = 0.9 \times 0.5d^2 = 0.45d^2$

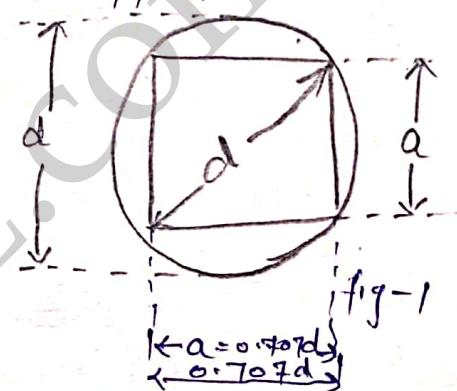


fig-1

$$d^2 = a^2 + a^2$$

$$d^2 = 2a^2 \Rightarrow d = \sqrt{2}a$$

$$d = \sqrt{2}a \Rightarrow a = \frac{d}{\sqrt{2}}$$

$$\text{Area of circle} = \pi r^2 = \pi \left(\frac{d}{2}\right)^2 = \frac{\pi}{4}d^2$$

$$\text{Ratio} = \frac{\text{Net core area}}{\text{area of circumscribing circle}} = \frac{0.45d^2}{\frac{\pi}{4}d^2} = 0.58$$

$$\text{Ratio} = \frac{\text{Gross core area}}{\text{area of circumscribing circle}} = \frac{0.5d^2}{\frac{\pi}{4}d^2} = 0.64$$

2. STEPPED CORES: →

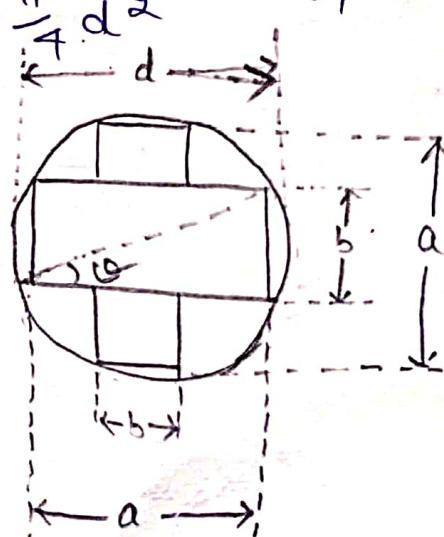
fig-2 shows a two stepped or cruciform core. The dimension of the two steps to give maximum area for a given diameter are determined as below

Gross Core area $A_{gi} = ab + b(a-b)$
 $= ab + ab - b^2 = 2ab - b^2$

Now, $a = d \cos \theta$ and $b = d \sin \theta$

$$A_{gi} = 2(d \cos \theta \cdot d \sin \theta) - d^2 \sin^2 \theta = 2d^2 \sin \theta \cos \theta - d^2 \sin^2 \theta$$

$(2 \sin \theta \cos \theta = \sin 2\theta)$



Differentiating $\epsilon_{pm}^{(2)}$ w.r.t. θ

29(b)

$$d^2(\alpha \cos \theta - \beta \sin \theta) / d\theta^2$$

$$\frac{dA_{gi}}{d\theta} = d^2 (\alpha \cos^2 \theta - \beta \sin^2 \theta)$$

$$\text{Equating } \frac{dA_{gi}}{d\theta} = 0 \Rightarrow \tan \theta = \alpha/\beta \Rightarrow \theta = 31^\circ 45'$$

$$\text{Therefore } a = d \cos 31^\circ 45' = 0.85d \quad b = d \sin 31^\circ 45' = 0.53d$$

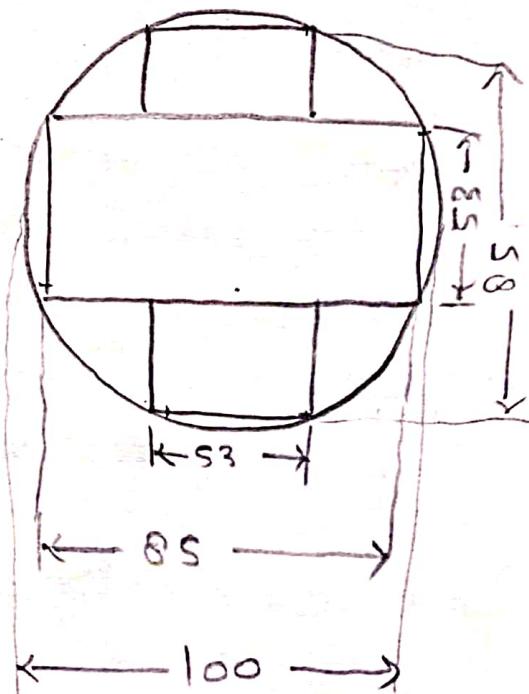
$$\therefore \text{Gross core area, } A_{gi} = \alpha ab - b^2 = \alpha(0.85d \cdot 0.53d) - (0.53d)^2$$

$$\text{Net core area} = 0.618d^2$$

$$A_i = k_s A_{gi} = 0.9 \times 0.618d^2 = 0.56d^2$$

$$\text{Ratio} = \frac{\text{Net core area}}{\text{area of circumscribing circle}} = \frac{0.56d^2}{(\frac{\pi}{4})d^2} = 0.71$$

$$\text{and } \frac{\text{Gross core area}}{\text{area of circumscribing circle}} = \frac{0.618d^2}{(\frac{\pi}{4})d^2} = 0.79$$



Selection of Design Constants:

Much of the design depends upon proper selection of design constants, flux density B_m , current density S , and window space factor K_w .

- * Choice of flux density, B_m : $\frac{1}{A_i} \propto B_m$
from voltage equation as well as output equation [$\Phi = \emptyset \cdot A_i B_m K_w S \times 10^{-3}$] indicates that, if we choose higher value of flux density B_m , the core area reduces. This will reduce the diameter of circumcircle thereby reducing the length of mean turn. Therefore there is a saving in cost of iron and copper. Also higher value of flux density increases the iron losses which will require the elaborate / more cooling arrangement.

The permissible value of flux densities are

Distn \Rightarrow 1.1 to 1.4 kWe/m^2

Power \Rightarrow 1.2 to 1.5 kWe/m^2

- * Choice of Current Density, S :

A higher value of current density will affect heating and efficiency. A higher value of current density will produce excessive temperature rise and insulation may be damaged due to intense local heating. The choice of current density affects the I^2R loss and efficiency hence also the load at which maximum efficiency occurs. The permissible value of current density are:-

① for standard Distn & small Power \Rightarrow

$$S = 1.5 \text{ to } 2.6 \text{ A/mm}^2 \text{ for large P.T}$$

② for std. & large power \Rightarrow $S = 5.4 \text{ to } 6.3 \text{ A/mm}^2$

Distn max - 3-5 A/mm²

* Choice of Window Space Factor - k_w

The window space factor, k_w , is the ratio of copper area in window to the total window area. All the total window area include the copper areas, the insulation, and air or oil spaces. The amount of copper and insulation used depend upon the kVA capacity and voltage rating resp. The choice of k_w depends upon kVA capacity and voltage rating. The following empirical relation:- for k_w

$$(i) \quad k_w = \frac{10}{30 + kV} \quad \text{b/w } 50 - 200 \text{ kVA}$$

$$(ii) \quad k_w = \frac{12}{30 + kV} \quad \text{for about } 1000 \text{ kVA}$$

$$(iii) \quad k_w = \frac{8}{30 + kV} \quad \text{for abt } 10 \text{ kVA} \\ (0.3 \text{ to } 0.7)$$

where kV is voltage in KV (kilo-volt of high voltage endg.

eg. Design a 250kVA, 2000/400V, 50Hz, 1-ph core type
GEN-35

Choice of Window Dimensions :-

The window area A_w , composed of two dimensions window height h_w and window width w_w . $A_w = h_w \times w_w$

→ Too narrow a window - gives increased height of wind. making wedges long and thin - In this case the dist. b/w adjac. Limb/wedges is less, the leakage resistance also reduces.

→ On the other hand if we choose lower height - window width (w_w) rises giving red dist b/w adjac. Limb/wedges. This will leads to increased leakage resistance.

Depending upon the suitability of window and desirability of leakage resistance, the height and width of window can be adjusted: the ratio generally lies b/w

$$\frac{h_w}{w_w} = 2 \text{ to } 4 \quad \text{the most usual value is 3.}$$

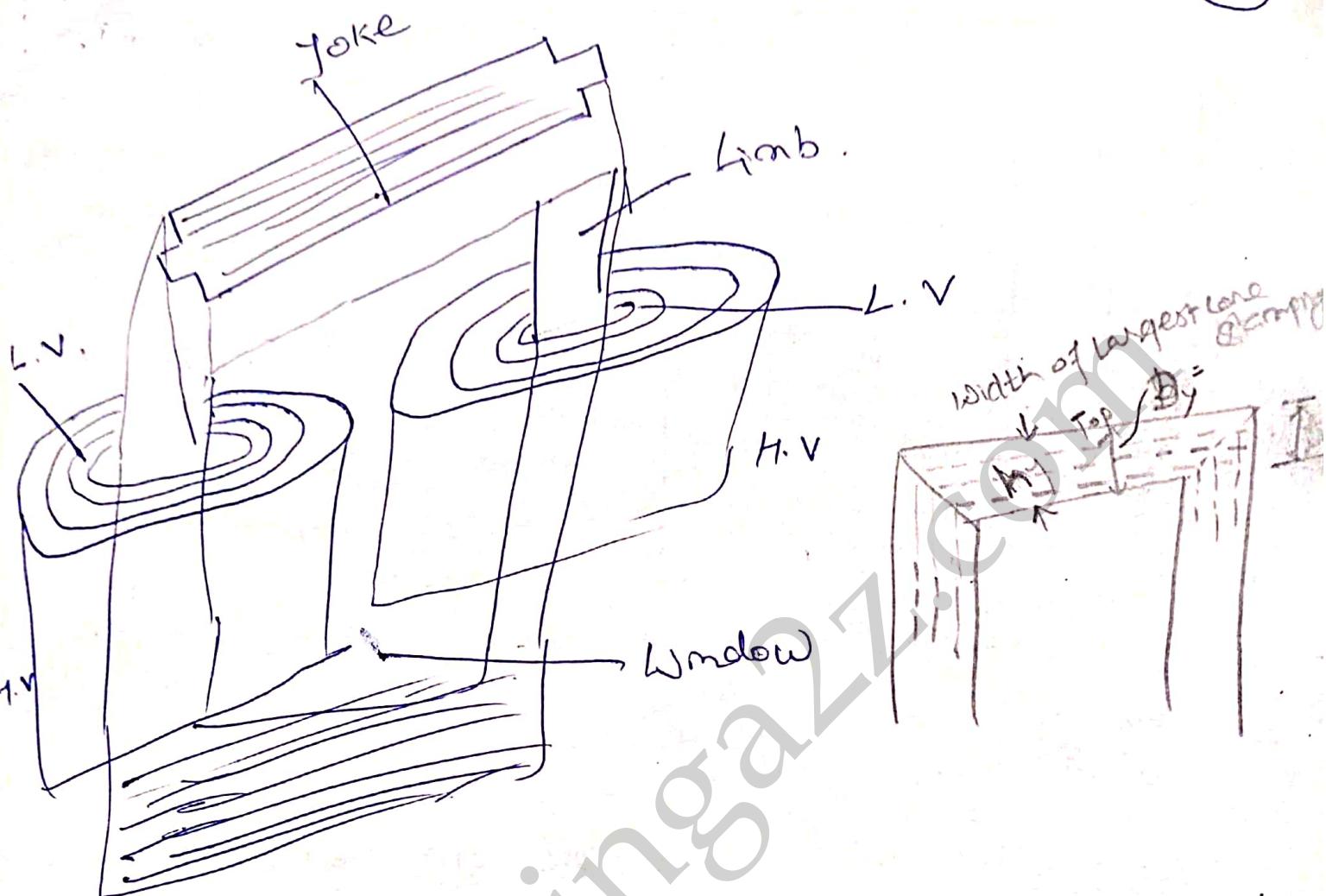
and $A_w = h_w \times w_w$ thus h_w and w_w can be calculated.

11 Dec 2020

* YOKE DIMENSION / DESIGN OF YOKE: $\frac{1}{A_i} \propto B_m$

No windows comes over the yoke portion, therefore there is possibility of reducing the iron loss by increasing the cross section area of yoke - how → reducing the flux density at yoke portion. Generally yoke area is taken as 15 to 25 percent higher than that of limb section.

The yoke section is most normally rectangular but it may be two stepped also. Considering the yoke section to be rectangular



Let
 A_y = Area of yoke, D_y = Depth of yoke
 H_y = Height of yoke

$$A_y = D_y \times H_y \quad \text{--- (1)}$$

Now, depth of yoke has to be same as the depth of core (as already obtained while designing the core section)

i.e., depth of yoke = Depth of core
= width of largest core stamping = $D_y = a$

and

$$A_y = (1.15 \text{ to } 1.25) A_{gi} \quad \text{--- (2)}$$

$$(1) \Rightarrow A_y = D_y \times H_y \quad \text{so } H_y = \frac{A_y}{D_y}$$

from depth of core

\hookrightarrow Thus H_y could be calculated.

OVERALL CORE DIMENSIONS:

(34)

After Designing the core & yoke section and also the window dimension, the overall dimensions can be calculated.

① for 1-phase (core) type t/2 : →

$$D = W_w + d$$

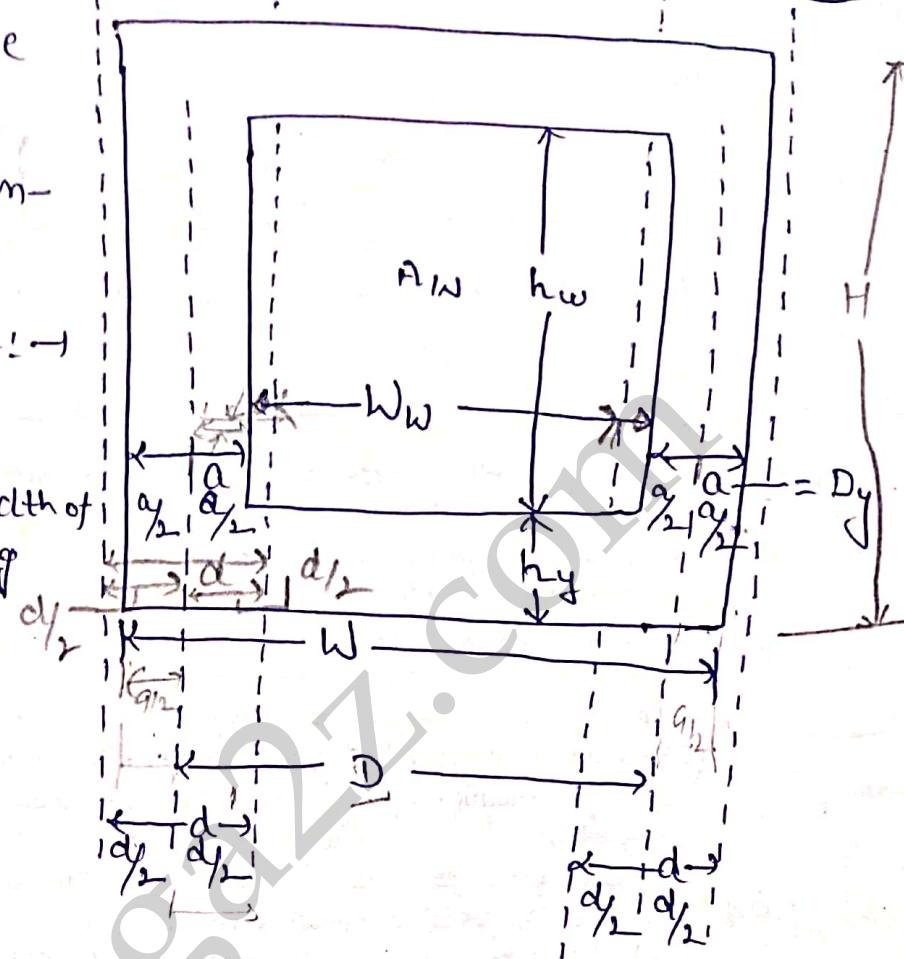
$D_y = a$ = Depth of yoke = width of longest core stamping

Let H = overall height

W = " width

$$H = h_w + 2h_y$$

$$W = D + a$$



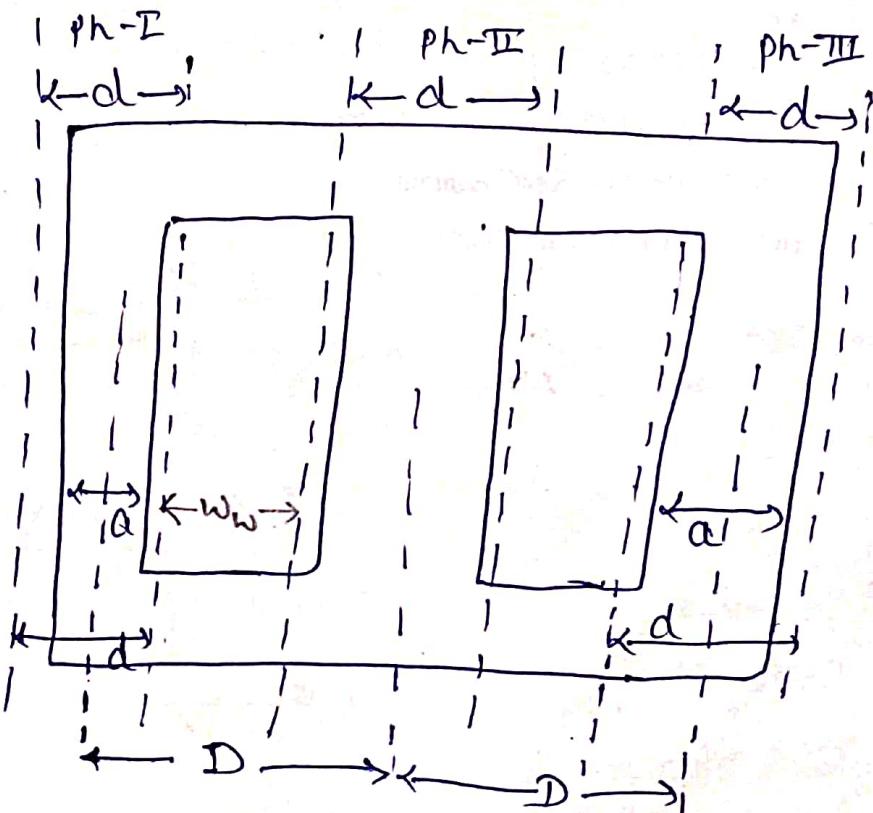
for 3-phase (core) type t/2 : →

$$D = W_w + d$$

$$D_y = a$$

$$H = h_w + 2h_y$$

$$W = 2D + a$$



Soln:- ① CORE

$A_{id,a}$

a = width of Largest
core stamping =
Depth of yoke

d = diam. of circumcircle
around the core

$$A_w = ? h_w \times w_w$$

$$\text{Q} = 2.22 f B_m A_i A_w K_w S \times 10^{-3}$$

Voltage per phase

$$\frac{\Sigma E_t}{T} = 4.44 f \phi_m = 4.44 f B_m A_i$$

$$\text{I } A_i = \frac{15}{4.44 \times 50 \times 1.25} = 0.05405 \text{ m}^2$$

$$\text{II } A_w = \frac{250}{2.22 \times 50 \times 1.25 \times 0.05405 \times 0.3 \times 2.75} \times 10^6 \times 10^{-3}$$

$$A_w = 0.04040 \text{ m}^2$$

$$A_w = ? h_w \times w_w = 0.04040 \text{ m}^2$$

$$\frac{h_w}{w_w} = 3 \Rightarrow \begin{cases} h_w = 3 w_w \\ h_w \times w_w = 0.04040 \\ 3 w_w^2 = 0.04040 \end{cases}$$

$$\begin{cases} w_w = 0.116 \text{ m} \\ h_w = 0.348 \text{ m} \end{cases}$$

② YOKE

Assuming 3-stepped section of yoke

Height of yoke, $h_y = a$ = Depth
of yoke, $d_y = a$
✓ overall height

$$H = h_w + d_y$$

$$= 0.348 + 2(0.27) = 0.888 \text{ m}$$

$$\text{overall width } W = D + a$$

$$\Rightarrow \text{Q} = W_w + d + a$$

$$= 0.116 + 0.3 + 0.27 = 0.686 \text{ m}$$

Assuming 3-stepped core section

$$A_i = 0.6 d^2$$

$$0.05405 = 0.6 d^2 \Rightarrow d = 0.30 \text{ m}$$

Now a = width of Largest stamping
 $a = 0.9 d = 0.27 \text{ m}$

Example: -

~ standard.

3-stepped 35

Design a 250kVA, 2000/400V, 50Hz, 1-phase, core type oil immersed, self cooled, power $\frac{1}{2}$ with full data. 3-stepped core needed.

Induced emf per turn = 15V, current den = $2.75A/mm^2$

Max. flux den in core = $1.25 Wb/m^2$, window space factor (k_w) = 0.3

Window proportion $\frac{\text{Height}}{\text{Width}} = 3$

Determine the main dimension of core and yoke

Soln:-

Voltage per turn = $\frac{E_t}{T} = \varepsilon_t = 4.44 f \phi_m = 4.44 f B_m A_i$

$$A_i = \frac{E_t}{4.44 f B_m} = \frac{15}{4.44 \times 50 \times 1.25} = 0.05405 m^2$$

Assuming a 3-stepped core section

$$A_i = 0.6d^2 \Rightarrow 0.05405 = 0.6d^2 \Rightarrow d = 0.30m$$

The width of largest stamping, $a = 0.9d$

Also depth of yoke

$$= 0.9 \times 0.3 = 0.27m \\ = 0.27m$$

KVA o/p of 1-ph $\frac{1}{2}$ is

$$\Phi = 2.22 f B_m A_i K_w A_w S \times 10^{-3}$$

$$250 = 2.22 \times 50 \times 1.25 \times 0.05405 \times A_w \times 0.3 \times 2.75 \times 10^6 \\ A_w = 0.04040 m^2 \times 10^{-3}$$

$$\text{Now } \frac{h_w}{w_w} = 3, \quad h_w \times w_w = 0.04040 m^2$$